

Off-line Power Supply Controller

FEATURES

- Transformerless Off-line
 Applications
- Ideal Primary-side Bias Supply
- Efficient BiCMOS Design
- Wide Input Range
- Fixed or Adjustable Low Voltage Output
- Uses Low Cost SMD Inductors
- Short Circuit Protected
- Optional Isolation Capability

DESCRIPTION

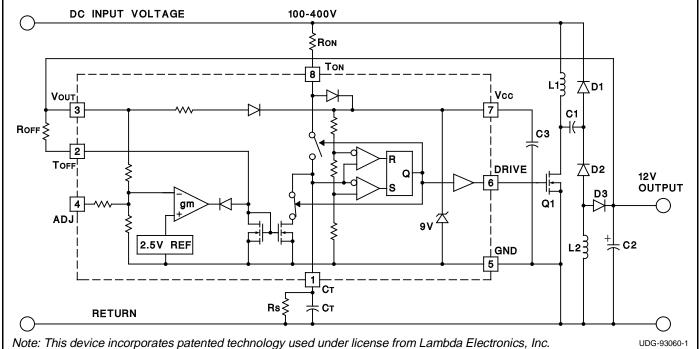
The UCC1889 controller is optimized for use as an off-line, low power, low voltage, regulated bias supply. The unique circuit topology utilized in this device can be visualized as two cascaded flyback converters, each operating in the discontinuous mode, and both driven from a single external power switch. The significant benefit of this approach is the ability to achieve voltage conversion ratios of 400V to 12V with no transformer and low internal losses.

The control algorithm utilized by the UCC1889 is to force the switch on time to be inversely proportional to the input line voltage while the switch off time is made inversely proportional to the output voltage. This action is automatically controlled by an internal feedback loop and reference. The cascaded configuration allows a voltage conversion from 400V to 12V to be achieved with a switch duty cycle greater than 10%. This topology also offers inherent short circuit protection since as the output voltage falls to zero, the switch off time approaches infinity.

The output voltage can be easily set to 12V or 18V. Moreover, it can be programmed for other output voltages less than 18V with a few additional components. An isolated version can be achieved with this topology as described further in Unitrode Application Note U-149.

OPERATION

With reference to the application diagram below, when input voltage is first applied, the RON current into TON is directed to VCC where it charges the external capacitor, C3, connected to VCC. As voltage builds on VCC, an internal undervoltage lockout holds the circuit off and the output at DRIVE low until VCC reaches 8.4V. At this time, DRIVE goes high turning on the power switch, Q1, and redirecting the current into TON to the timing capacitor, CT. CT charges to a fixed threshold with a current ICHG=0.8 • (VIN - 4.5V)/RON. Since DRIVE will only be high for as long as CT charges, the power switch on time will be inversely proportional to line voltage. This provides a constant line voltage-switch on time product.



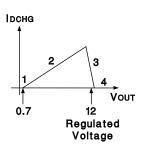
TYPICAL APPLICATION



OPERATION (cont.)

At the end of the on time, Q1 is turned off and the RON current into TON is again diverted to Vcc. Thus the current through RON, which charges CT during the on time, contributes to supplying control power during the off time.

The power switch off time is controlled by the discharge of CT which, in turn, is programmed by the regulated output voltage. The relationship between CT discharge current, IDCHG, and output voltage is illustrated as follows:



- When VOUT = 0, the off time is infinite. This feature provides inherent short circuit protection. However, to ensure output voltage startup when the output is not a short, a high value resistor, Rs, is placed in parallel with CT to establish a minimum switching frequency.
- 2. As VOUT rises above approximately 0.7V to its regulated value, IDCHG is defined by ROFF, and therefore is equal to:

IDCHG = (VOUT - 0.7V) / ROFF

As V_{OUT} increases, IDCHG increases resulting in the reduction of off time. The frequency of operation increases and V_{OUT} rises quickly to its regulated value.

- 3. In this region, a transconductance amplifier reduces IDCHG in order to maintain VOUT in regulation.
- 4. If VOUT should rise above its regulation range, IDCHG falls to zero and the circuit returns to the minimum frequency established by Rs and CT.

The range of switching frequencies is established by RON, ROFF, Rs, and CT as follows:

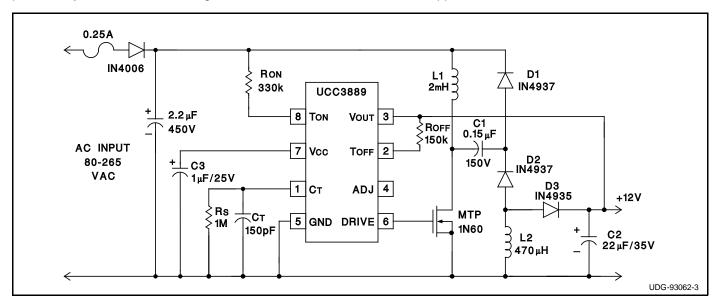
Frequency = 1/(TON + TOFF)TON = RON • CT • 4.6 V/(VIN - 4.5V) TOFF (max) = 1.4 • Rs • CT Regions 1 and 4

TOFF = ROFF • CT • 3.7V /(VOUT - 0.7V) Region 2, excluding the effects of Rs which have a minimal impact on TOFF.

The above equations assume that VCC equals 9V. The voltage at TON increases from approximately 2.5V to 6.5V while CT is charging. To take this into account, VIN is adjusted by 4.5V in the calculation of TON. The voltage at TOFF is approximately 0.7V.

DESIGN EXAMPLE

The UCC3889 regulates a 12 volt, 1 Watt nonisolated DC output from AC inputs between 80 and 265 volts. In this example, the IC is programmed to deliver a maximum on time gate drive pulse width of 2.4 microseconds which occurs at 80 VAC. The corresponding switching frequency is approximately 100kHz at low line, and overall efficiency is approximately 50%. Additional design information is available in Unitrode Application Note U-149.

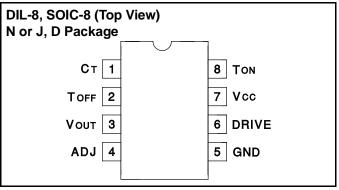


UCC1889 UCC2889 UCC3889

ABSOLUTE MAXIMUM RATINGS

Icc 5mA
Current into Ton Pin
Voltage on Vout Pin 20V
Current into TOFF Pin
Storage Temperature
Note: Unless otherwise indicated, voltages are referenced to
ground and currents are positive into, negative out of, the speci-
fied terminals.

CONNECTION DIAGRAM



ELECTRICAL CHARACTERISTICS Unless otherwise stated, these specifications hold for TA = 0°C to 70°C for the UCC3889, -40°C to +85°C for the UCC2889, and -55°C to +125°C for the UCC1889. No load at DRIVE pin (CLOAD=0).

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
General		-			-
Vcc Zener Voltage	Icc < 1.5mA	8.6	9.0	9.3	V
Startup Current	Vout = 0		150	250	μΑ
Operating Current I(VOUT)	Vout = 11V, F = 150kHz		1.2	2.5	mA
Under-Voltage-Lockout	•				
Start Threshold	Vout = 0	8.0	8.4	8.8	V
Minimum Operating Voltage after Start	Vout = 0	6.0	6.3	6.6	V
Hysteresis	Vout = 0	1.8			V
Oscillator		-			-
Amplitude	Vcc = 9V	3.5	3.7	3.9	V
CT to DRIVE high Propagation Delay	Overdrive = 0.2V		100	200	ns
CT to DRIVE low Propagation Delay	Overdrive = 0.2V		50	100	ns
Driver		-			-
VOL	I = 20mA, Vcc = 9V		0.15	0.4	V
	I = 100mA, Vcc = 9V		0.7	1.8	V
VOH	I = -20mA, Vcc = 9V	8.5	8.8		V
	I = -100mA, Vcc = 9V	6.1	7.8		V
Rise Time	CLOAD = 1nF		35	70	ns
Fall Time	CLOAD = 1nF		30	60	ns
Line Voltage Detection					
Charge Coefficient: Існс / І(Тол)	VCT = 3V, DRIVE = High, I(TON) = 1mA	0.73	0.79	0.85	
Minimum Line Voltage for Fault	Ron = 330k	60	80	100	V
Minimum Current I(Ton) for Fault	Ron = 330k		220		μA
On Time During Fault	CT = 150pF, $VLINE = Min - 1V$		2		μs
Oscillator Restart Delay after Fault			0.5		ms
Vout Error Amp					
VOUT Regulated 12V (ADJ Open)	VCC = 9V, IDCHG = $I(TOFF)/2$	11.2	11.9	12.8	V
VOUT Regulated 18V (ADJ = 0V)	VCC = 9V, IDCHG = I(TOFF)/2	16.5	17.5	19.5	V
Discharge Ratio: IDCHG / I(TOFF)	I(Toff) = 50μA	0.93	1.00	1.07	
Voltage at TOFF	I(Toff) = 50μA	0.6	0.95	1.3	V
Regulation gm (Note 1)	Мах Ідснд = 50µА		1.0		mA/V
	Max IDCHG = 125µА	0.8	1.7	2.9	mA/V

Note 1: gm is defined as $\frac{\Delta \text{IDCHG}}{\Delta \text{VOUT}}$ for the values of VOUT when VOUT is in regulation. The two points used to calculate gm are for IDCHG at 65% and 35% of its maximum value.

PIN DESCRIPTIONS

ADJ: The ADJ pin is used to provide a 12V or an 18V regulated supply without additional external components. To select the 12V option, ADJ pin is left open. To select the 18V option, ADJ pin must be grounded. For other output voltages less than 18V, a resistor divider between VOUT, ADJ and GND is needed. Note, however, that for output voltages less than Vcc, the device needs additional bootstrapping to Vcc from an external source such as the line voltage. If so, precautions must be taken to ensure that total lcc does not exceed 5mA.

CT (timing capacitor): The signal voltage across CT has a peak-to-peak swing of 3.7V for 9V Vcc. As the voltage on CT crosses the oscillator upper threshold, DRIVE goes low. As the voltage on CT crosses the oscillator lower threshold, DRIVE goes high.

DRIVE: This output is a CMOS stage capable of sinking 200mA peak and sourcing 150mA peak. The output voltage swing is 0 to Vcc.

GND (chip ground): All voltages are measured with respect to GND.

TOFF (regulated output control): TOFF sets the discharge current of the timing capacitor through an external

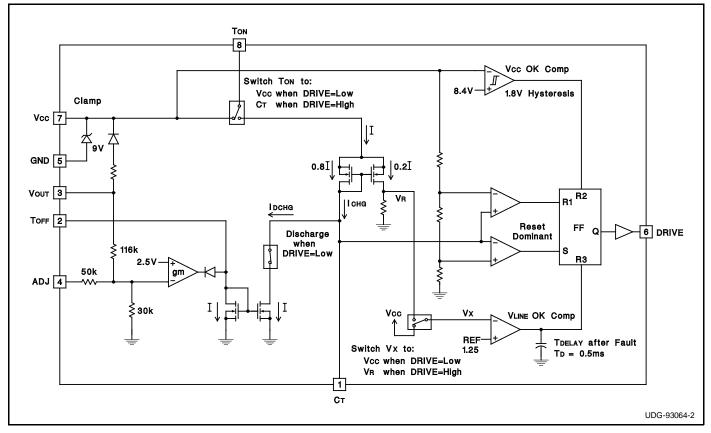
resistor connected between VOUT and TOFF.

TON (line voltage control): TON serves three functions. When CT is discharging (off time), the current through TON is routed to VCC. When CT is charging (on time), the current through TON is split 80% to set the CT charge time and 20% to sense minimum line voltage which occurs for a TON current of 220μ A. For a minimum line voltage of 80V, RON is 330kΩ.

The CT voltage slightly affects the value of the charge current during the on time. During this time, the voltage at the TON pin increases from approximately 2.5V to 6.5V.

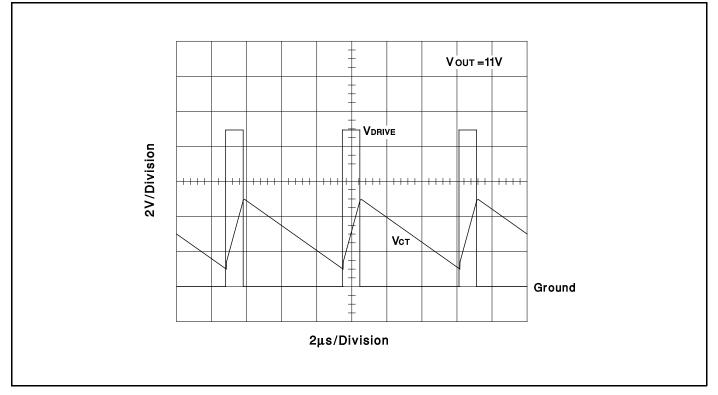
Vcc (chip supply voltage): The supply voltage of the device at pin Vcc is internally clamped at 9V. Normally, Vcc is not directly powered from an external voltage source such as the line voltage. In the event that Vcc is directly connected to a voltage source for additional bootstrapping, precautions must be taken to ensure that total lcc does not exceed 5mA.

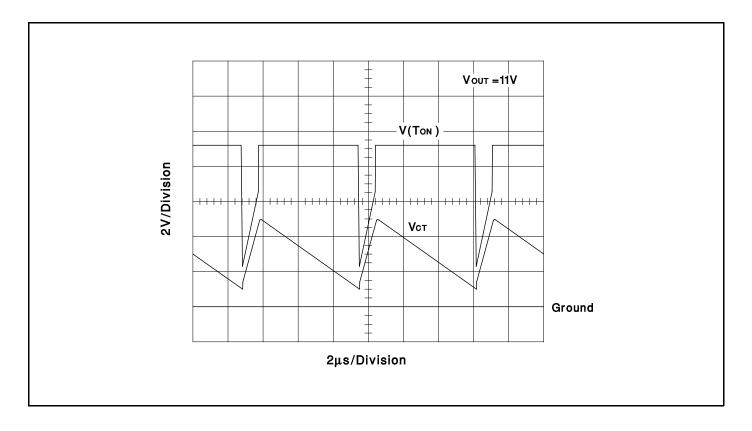
VOUT (regulated output): The VOUT pin is directly connected to the power supply output voltage. When VOUT is greater than VCC, VOUT bootstraps VCC.



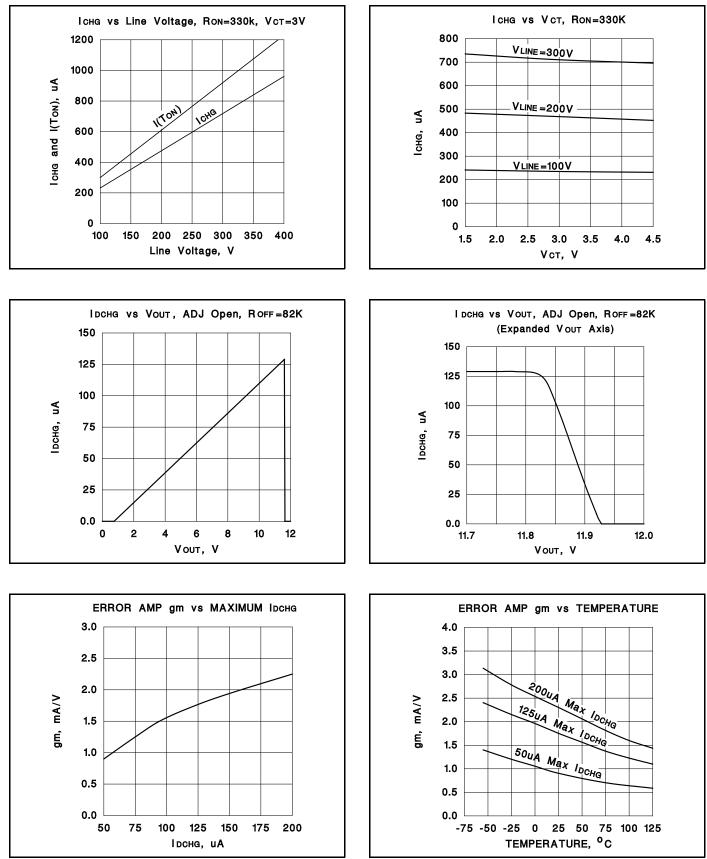
BLOCK DIAGRAM

TYPICAL WAVEFORMS





UCC1889 UCC2889 UCC3889



UNITRODE INTEGRATED CIRCUITS 7 CONTINENTAL BLVD. • MERRIMACK, NH 03054 TEL. 603-424-2410 • FAX 603-424-3460

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Packag Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
UCC2889D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC2889DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC2889DTR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC2889DTR/81361G4	PREVIEW	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC2889DTRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC2889N	ACTIVE	PDIP	Р	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
UCC2889NG4	ACTIVE	PDIP	Р	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
UCC3889D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC3889DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC3889DTR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC3889DTRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
UCC3889N	ACTIVE	PDIP	Р	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
UCC3889NG4	ACTIVE	PDIP	Р	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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PACKAGE OPTION ADDENDUM

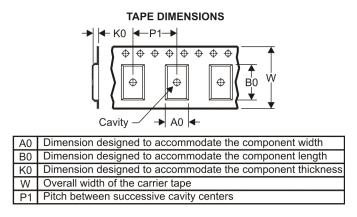


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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

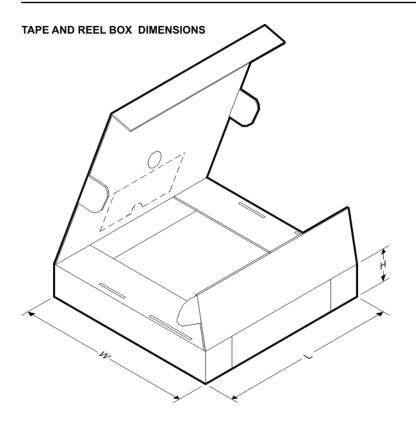


*/	All dimensions are nominal												
	Device		Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
	UCC2889DTR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
	UCC3889DTR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



PACKAGE MATERIALS INFORMATION

29-Jul-2008

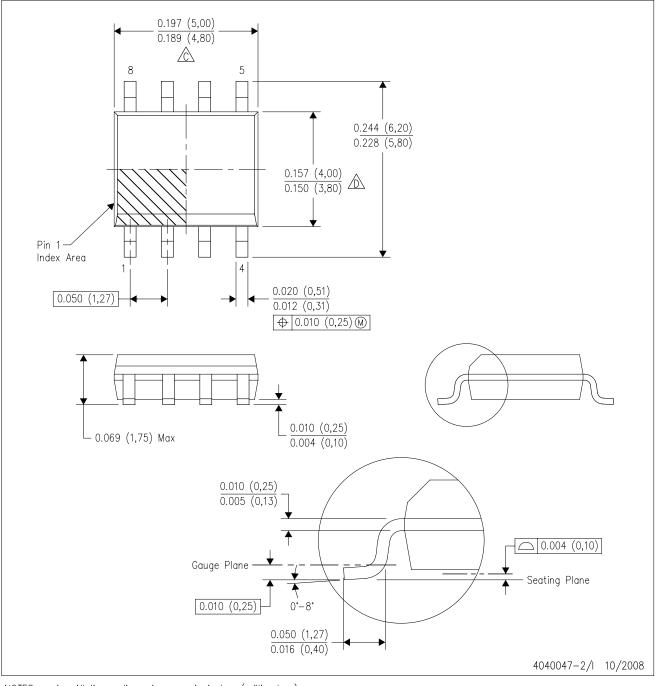


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
UCC2889DTR	SOIC	D	8	2500	346.0	346.0	29.0
UCC3889DTR	SOIC	D	8	2500	346.0	346.0	29.0

D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.

Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.

E. Reference JEDEC MS-012 variation AA.



MECHANICAL DATA

MPDI001A - JANUARY 1995 - REVISED JUNE 1999



- NOTES: A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001

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